

**FERROXCUBE**

*Engineering  
Bulletin*

FERROXCUBE CORPORATION of AMERICA, SAUGERTIES, NEW YORK

FEBRUARY, 1958

No. FC-5119

# MAGNETIC

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# DEFINITIONS

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**FIRST IN FERRITES**

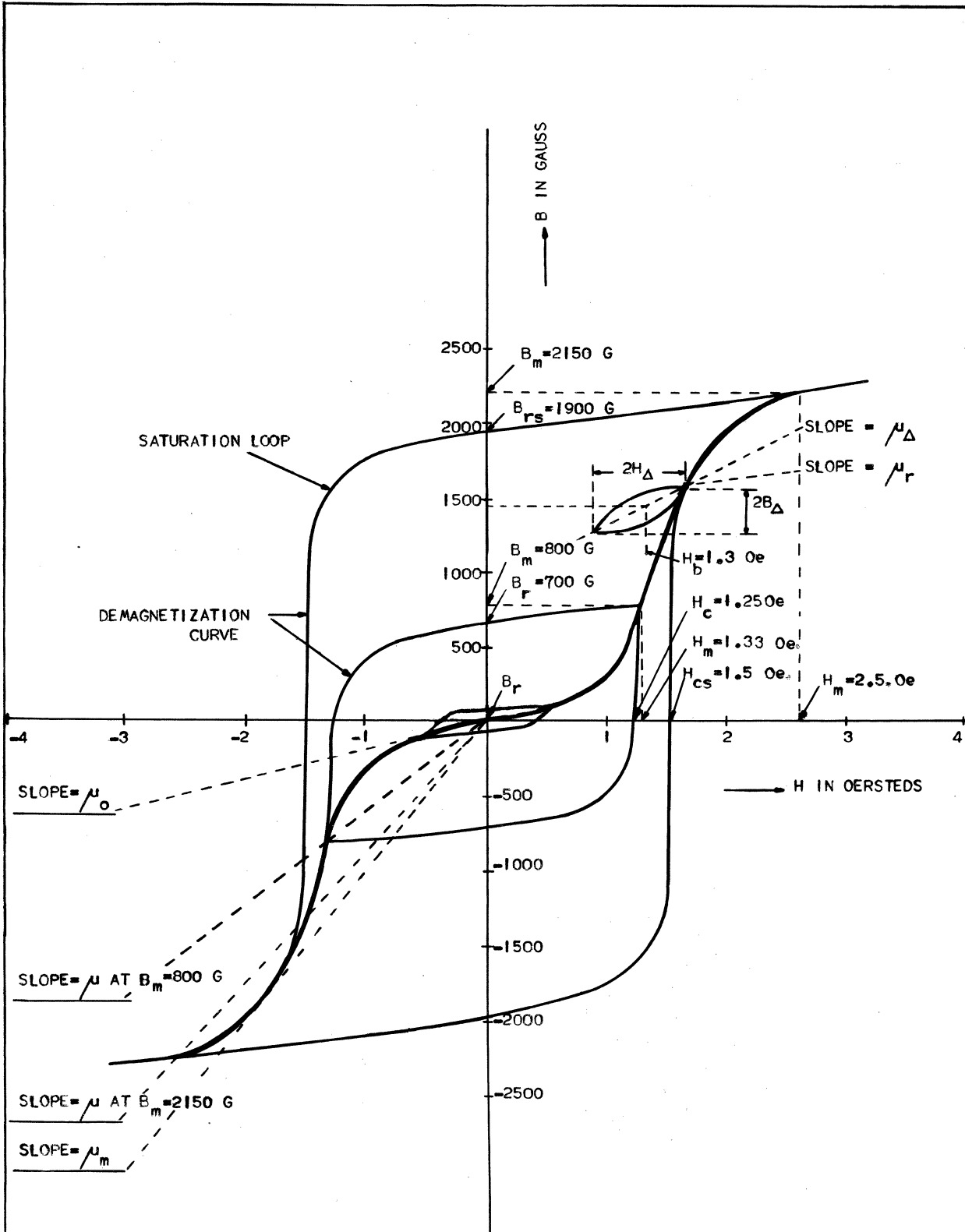


FIGURE 1

HYSTERESIS CURVE OF A SQUARE LOOP MATERIAL

NOTES:  $B_s + H$  DOES NOT APPEAR  
IN THIS DRAWING. IT  
WILL BE ABOUT 2500 G

— = VIRGINAL CURVE

1. FERROMAGNETIC MATERIAL

**Ferromagnetic Material**

A material which, like iron, generally exhibits hysteresis, i.e. the induction fails to follow the magnetic field, it lags: See fig. 1.

**Ferrite**

A ceramic material that is sintered at about 1300°C. It mainly consists of iron oxide with a certain additive of, usually, two metal oxides. Most current are the Manganese-Zinc, Nickel-Zinc, Manganese-Magnesium and Manganese-Copper ferrites. These are called soft magnetic materials because of comparatively low coercive forces, say smaller than 20 oersteds (e.g. Ferroxcube). See fig. 1. Barium ferrites have a very high coercive force (say 1000 Oe) and are called hard magnetic materials (e.g. Magnadur). See fig. 11.

2. FLUX

SYMBOL	UNIT	TERM	DESCRIPTION
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$\phi$  maxwells (Magnetic Flux)

The cause of the electromotive force (voltage) induced in the secondary windings of a core (see fig. 2), magnetized by an A.C. current in the primary windings. In the c.g.s. system:

$$e_{ind} = -N_2 \times \frac{d\phi}{dt} \times 10^{-8}$$

where  $e_{ind}$  = induced e.m.f. in volts;  
 $N_2$  = number of secondary windings;  
 $\frac{d\phi}{dt}$  = change of flux per second in maxwells/sec.

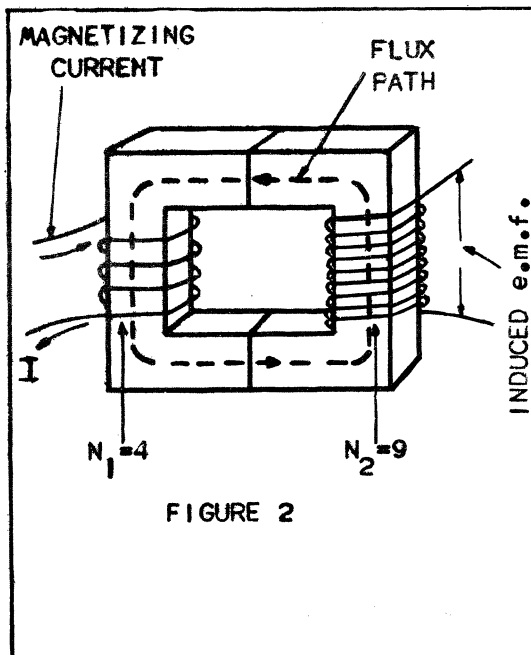
$\phi$  itself is in maxwells.

Note: If  $\phi$  is an alternating flux with frequency  $f$ , then, the more times per second that  $\phi$  reverses its direction,  $e_{ind}$  will be proportionally larger. In case of a sinus wave:

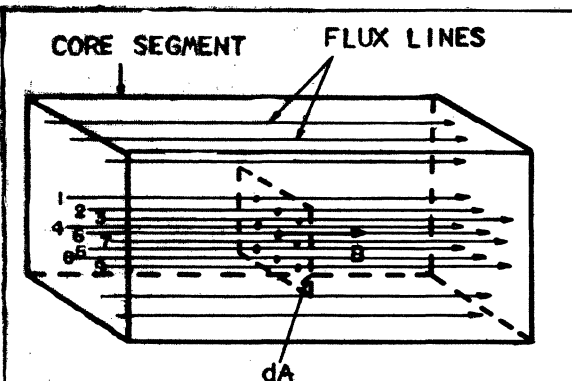
$$\hat{e}_{ind} = N_2 \times 2\pi f \times \hat{\phi} \times 10^{-8},$$

where:  $\hat{e}_{ind}$  and  $\hat{\phi}$  indicate the peak-values of  $e_{ind}$  and  $\phi$ .

$f$  = frequency in cycles per sec.



3. INDUCTION

SYMBOL	UNIT	TERM	DESCRIPTION
B	gauss	(Magnetic) Induction (Magnetic) Flux Density	<p>Magnetic flux per unit area in an area element at right angles to the direction of the flux lines. In the c.g.s. system:</p> $B = \frac{d\phi}{dA}$ <p>Here <math>d\phi</math> denotes a very small fraction of the total flux passing through a very small fraction of the area at a right angle: theoretically B changes from place to place in a material. In practice, when B is constant across the area, we use:</p> $B = \frac{\phi}{A}$ <p>Where: B = induction in gaussess  <math>\phi</math> = flux in maxwells                      A = area in cm<sup>2</sup>.</p> <p>See fig. 3.</p>
<div style="border: 1px solid black; padding: 10px;">  <p>CORE SEGMENT      FLUX LINES</p> <p>OF ALL FLUX-LINES THE NUMBERS 1 THROUGH 9, TOGETHER AMOUNTING TO A FLUX <math>d\phi</math>, PASS THROUGH <math>dA</math>.</p> <math display="block">B = \frac{d\phi}{dA}</math> <p>FIGURE 3</p> </div>			
$B_r$	gauss	Residual Induction	Induction that remains, when the magnetizing force $H = 0$ . Compare with $B_d$ and see fig. 1.
$B_d$	gauss	Remanent Induction Remanence	Induction that remains in a magnetic circuit of any shape after removal of an applied magnetomotive force, $\mathcal{H}$ . In a permanent magnet with an air gap the magnetomotive force = 0, but the magnetic field in the magnet itself, $H$ , is not zero, its value depending on the form of its hysteresis loop (energy product, see page 19) and width of the air-gap.
$B_i$	gauss	Intrinsic Induction	<p>The induction in the magnetic material minus the induction in air at the existing magnetic field, <math>H</math>. The induction in air = <math>\mu_{air} H</math>. In c.g.s. units:</p> $\mu_{air} = 1, \text{ therefore } B_i = B - H. \text{ (See } \mu_i \text{)}$

SYMBOL	UNIT	TERM	DESCRIPTION
$B_{max}$ or $B_m$	gauss	Maximum Induction	Maximum instantaneous value of the induction on a hysteresis loop. It can be measured by means of an integrating network, which converts the induced A.C. e.m.f. into an A.C. voltage $V$ proportional to $B$ . The peak value of $V$ yields: $\hat{B} = \frac{RC \hat{V} \times 10^8}{N_2 A} \text{ gauss}$ where $RC$ = time constant of the integrator in sec., $\hat{V}$ = the peak value of the integrated e.m.f. in volts, $N_2$ = the number of secondary windings, $A$ = the area in $cm^2$ . See fig. 1 and 2.
$B_{\Delta}$	gauss	Incremental Induction	Amplitude (= peak value) of the varying part of the induction, when the core is operated with a biasing magnetizing force. See fig. 1.
$B_{rs}$	gauss	Retentivity	The maximum value of the residual induction, $B_r$ or the value of $B_r$ on the saturation loop. See fig. 1.
$B_{sat}$ or $B_s$	gauss	Saturation Induction	The maximum intrinsic induction possible in a material. It is usually reached at magnetic fields of thousands of oersteds, while at the point where the saturation loop is closed, $B_i$ is 10 to 30 percent lower than $B_s$ , and $H$ is roughly 10 oersteds. See fig. 1.

4. MAGNETIC FIELD

SYMBOL	UNIT	TERM	DESCRIPTION
$\mathcal{F}$	gilberts	Magnetomotive Force Magnetic Potential Difference	That which tends to produce a magnetic field. In magnetic testing it is most commonly produced by a current flowing through a coil of wire around the core. Its magnitude is proportional to the number of turns and the current. In the c.g.s. unit-system it is expressed in gilberts and thus defined by: $\mathcal{F} = 0.4 \pi N I,$ where $I$ = current in amperes Magnetomotive force also results from a magnetized body, e.g. a permanent magnet.

$H$	oersteds	Magnetizing Force Magnetic Field (Intensity)	Magnetomotive force per unit length. In the c.g.s. unit-system it is expressed in oersteds and is defined by $H = \frac{d\mathcal{F}}{dl}$ . Here $d\mathcal{F}$ denotes a very small change of $\mathcal{F}$ along a very small path length; generally $H$ changes from point to point in direction and intensity. $H$ has the same direction as $dl$ . In practice $H$ has at all points the same value in most applications. We can then use:
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VALUES OF MAGNETOMOTIVE FORCE

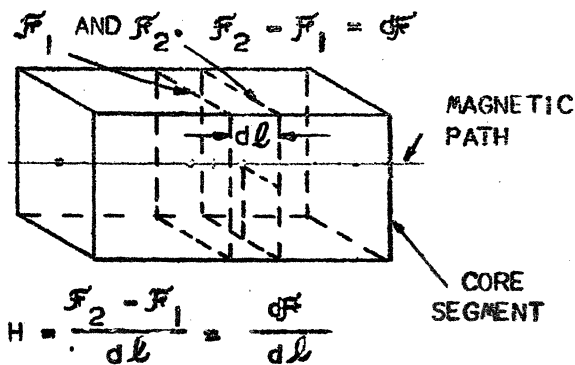


FIGURE 4

where  $H = \frac{\mathcal{F}}{l}$   
 $H$  = magnetizing force in oersteds,  
 $\mathcal{F}$  = magnetomotive force in gilberts,  
 $l$  = path length in cm.

See fig. 4.  
 In the case of a toroid,  
 $l = 2 \pi r$

where  $r$  = the average of inside and outside radii of the ring :

$$H = \frac{0.4 \pi N I}{2 \pi r} = \frac{0.2 N I}{r}$$

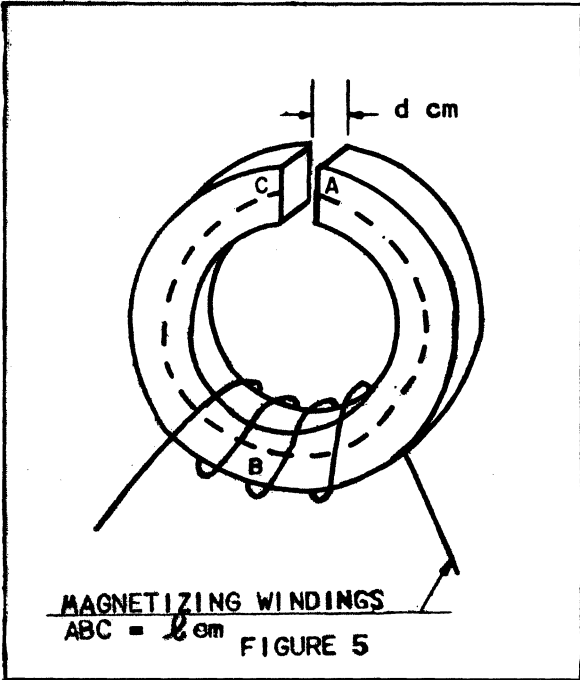
$H_b$	oersteds	Biassing (Magnetic) Field	A constant (D.C.) magnetizing force superposed on an alternating (A.C.) magnetizing force. This is realized by applying an extra D.C. current to the same magnetizing windings or to a special coil around the same sample. See fig. 1.
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SYMBOL	UNIT	TERM	DESCRIPTION
$H_c$	oersteds	Coercive Force	The magnetizing force required to bring the induction to zero; the larger the $H_c$ , the larger are the hysteresis losses. See fig. 1.
$H_{cs}$	oersteds	Coercivity	The maximum value of the coercive force, $H_c$ , or the value of $H_c$ on the saturation loop. See fig. 1.
$H_d$	oersteds	Demagnetizing Force	A magnetizing force applied in such a direction as to reduce the remanent induction, $B_d$ , in a magnetized body. Fig. 1, upper left, and fig. 11 give examples of a demagnetization curve.
$H_\Delta$	oersteds	Incremental Magnetizing Force	Amplitude of the varying part of the magnetizing force, when the core is operated with a biasing field, $H_b$ . See Fig. 1.
$H_{max}$ or $H_m$	oersteds	Maximum Magnetizing Force	Maximum instantaneous value of the magnetic field on a hysteresis loop. $H_m$ (and $B_m$ ) are found on the "tip" of the hysteresis loop.

5. PERMEABILITY

SYMBOL	UNIT	TERM	DESCRIPTION
$\mu$	numeric	Permeability $\mu$	<p>The slope of the line, which connects the "tip" of a hysteresis loop (<math>H_{max}</math>, <math>B_{max}</math>) with the origin, or <math>\mu = \frac{B_{max}}{H_{max}}</math> (usually written <math>\mu = \frac{B}{H}</math>). See Fig. 1.</p> <p>The "tips" of all possible loops that can be performed, depending on the amplitude of the applied magnetic field, <math>H_{max}</math> are situated on the so-called virginal curve. This curve is straight near the origin; its slope corresponds to the permeability at low values of H or initial permeability, <math>\mu_0</math>.</p> <p>Note 1: Depending on the method of measurement and the practical application, the <math>\mu</math> can also be defined as the R.M.S. value of the A.C. induction divided by the R.M.S. value of the A.C. magnetic field or <math>\mu = \frac{B_{R.M.S.}}{H_{R.M.S.}}</math>. Now the results are dependent among others on the wave shape of the magnetic field and the form of the hysteresis loop.</p> <p>Note 2: In such non-isotropic materials as Magnadur and Ferroplane, <math>\mu</math> depends on the direction of H with respect to the sample: there is a direction or a plane in the sample where <math>\mu</math> is maximal.</p>
$\mu_{max}$ or $\mu_m$	numeric	Maximum Permeability	<p>Maximum value of the permeability. It is the slope of the line which just touches the "knee" in the virginal curve. See Fig. 1.</p>
$\mu_0$	numeric	Initial Permeability	<p>The permeability at low inductions, say 10 gauss or less. See Fig. 1.</p>
$\mu_{\Delta}$	numeric	Incremental Permeability	<p>The ratio of the incremental induction to the incremental magnetizing force or <math>\mu_{\Delta} = \frac{B_{\Delta}}{H_{\Delta}}</math>. When <math>H_{\Delta}</math> and consequently <math>B_{\Delta}</math> become very small, then <math>\mu_{\Delta}</math> approaches a limiting value, the reversible permeability <math>\mu_{rev}</math>. (See Fig. 1) <math>\mu_{rev}</math> decreases from <math>\mu_{rev} = \mu_0</math> (<math>H_b = 0</math>) down to <math>\mu_{rev} = 1</math> when <math>H_b</math> is very high.</p>
$\mu_{rev}$	numeric	Reversible Permeability	



SYMBOL	UNIT	TERM	DESCRIPTION
$\mu_{eff}$	numeric	Effective Permeability	<p>The lowered value of <math>\mu</math> caused by interruptions in the flux path. If a magnetic circuit, e.g. a toroid, contains an air-gap, then the effective induction <math>B_{eff}</math> has a lower value than in the case without an air-gap at a given value of <math>H</math>. The effective permeability is defined as:</p> $\mu_{eff} = \frac{B_{eff}}{H}$ <p><math>\mu_{eff}</math> is always smaller than <math>\mu</math> and can be calculated as follows:</p> $\frac{\mu_{eff}}{\mu} = \frac{1}{1 + \mu (g/l)}$ <p>where:</p> <p><math>g</math> = width of the air-gap in cm,  <math>l</math> = magnetic path length of the core in cm. See figure 5.</p> <p>Gaps caused by dirt often cause difficulties when testing U cores. In practice they are designed into pot cores in order to reduce both the temperature coefficient of the pot core assembly and the effects of non-linearity.</p>
			
$\mu_i$	numeric	Intrinsic Permeability	<p>The ratio of the intrinsic induction to the corresponding magnetizing force:</p> $\mu_i = \frac{B_i}{H}$ <p>Because <math>B_i = B - H</math>, we see that <math>\mu_i = \frac{B-H}{H} = \mu - 1</math>. The use of <math>\mu_i</math> becomes important for materials with very low <math>\mu</math>, say <math>\mu = 10</math>. The <math>\mu_i</math> in this case is 9.</p>
$\mu'$	numeric	$\mu$ prime	<p>Real part of the so-called complex permeability. At low inductions it is identical with the initial permeability.</p>
$\mu''$	numeric	$\mu$ double prime	<p>This is the imaginary part of the complex permeability (<math>\mu' - j \mu''</math>). This is only a formal way of appreciating the core losses relative to the initial permeability <math>\mu_0</math> or <math>\mu'</math>.</p>

6. INDUCTANCE

SYMBOL	UNIT	TERM	DESCRIPTION
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L henries Self Induc-  
tance

1) Electrically defined by:

$$e_{ind} = -L \frac{dI}{dt},$$

where  $e_{ind}$  = induced e.m.f. across the coil windings in volts,

L = self inductance in henries,

$\frac{dI}{dt}$  = change of current per second in amperes per sec.

Measurement of L is based on this principle.

2) An equivalent definition is:

$$L = \frac{N\phi}{I} \times 10^{-8},$$

where  $N\phi$  = flux linkage in maxwell-turns,

I = magnetizing current in amperes,

L = self inductance in henries.

"Flux-linkage" means the total flux that is "caught" by the conductor as a whole (usually a coil around a core). If the wire is homogeneously wound around the whole circumference of a toroid, then N equals the total number of turns. With the help of the definition (2) we can derive for the toroid the formula:

$$3) L = \frac{0.4 \pi \mu_0 \times N^2 \times 10^{-8}}{l/A},$$

where  $\mu_0$  = initial permeability

N = number of turns

A = cross section of toroid in  $cm^2$

l = path length in cm (for a toroid about the average of inside and outside circumference).

At high inductions  $\mu$  cannot be considered a constant (fig. 1), hence L is then not well-defined.

SYMBOL	UNIT	TERM	DESCRIPTION
$L_o$	henries	Air Inductance	<p>Defined as <math>L</math> under (3), except that <math>\mu_o</math> is put = 1 (permeability of air). This provides a method of measuring <math>\mu_o</math>:</p> $L = \mu_o L_o,$ <p>or</p> $\mu_o = \frac{L}{L_o},$ <p>where <math>L</math> is measured and <math>L_o</math> is calculated from the physical dimensions of the core. The general formula for cores with non-uniform cross section is:</p> $L_o = \frac{0.4 \pi \times N^2 \times 10^{-8}}{\sum \ell/A}$ <p><math>\Sigma</math> or "sigma" denotes that one takes the sum of the <math>\ell/A</math>'s for the different parts of the magnetic circuit (e.g. E core, pot core).</p>

$L_m$ or $M$	henries	Mutual Inductance	<p>If we have a transformer (see fig. 2) with <math>N_1</math> primary windings and <math>N_2</math> secondary windings, then both coils are linked magnetically by the core. This property is characterized by the mutual inductance:</p> $L_m = \frac{N_2 \phi \times 10^8}{I_1} = \frac{N_1 \phi \times 10^8}{I_2}$ <p>Where:</p> <ul style="list-style-type: none"> <li><math>L_m</math> = mutual inductance in henries,</li> <li><math>N_2 \phi</math> (<math>N_1 \phi</math>) = flux linkage "caught" by the secondary (primary) windings in maxwell turns,</li> <li><math>I_1</math> (<math>I_2</math>) = current in the primary (secondary) windings in amperes.</li> </ul> <p>In an ideal transformer (no flux escaping through surrounding air)</p> $L_m^2 = L_1 \times L_2, \quad L_1 \text{ and } L_2 \text{ being the}$ <p>the individual self-inductances of the primary and secondary windings.</p>
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7. LOSSES

SYMBOL	UNIT	TERM	DESCRIPTION
$P_c$	watts	Core Loss	<p>Energy per second (= power) dissipated in a core when magnetized by an alternating magnetic field. The dissipated power is identical with the developed heat per second, power (heat per sec.) being in watts. <math>P_c = P_e + P_h + P_r</math></p> <p>See below.</p>
$P_e$	watts	Eddy Current	<p>Power dissipated due to currents circulating in the core as a result of e.m.f.'s in the material induced by the varying induction. <math>P_e</math> can be calculated from <math>P_e = \frac{\omega^2 B^2 \delta^2 V}{9} \times 10^{-14} \times C_c</math> watts</p> <p>where: <math>\omega</math> = angular frequency,  <math>B</math> = R.M.S. value of the induction in gauss,  <math>\delta</math> = smallest dimension in cm of core perpendicular to the flux,  <math>V</math> = volume of the core in <math>cm^3</math>,  <math>9</math> = resistivity in ohm cm,  <math>C_c</math> = shape constant, which is <math>1/\delta</math> for cores with a uniformly round or square cross-section.</p>
$P_h$	watts	Hysteresis Loss	<p>Power dissipated in a ferromagnetic core as a result of its hysteresis. It is proportional to the product of the area of the loop x frequency x volume of the core. <math>P_h</math> can be calculated from:</p> $P_h = \frac{fV}{4\pi} \times \text{loop area} \times 10^{-7} \text{ watts}$ <p>where: <math>f</math> = frequency in cycles/sec,  <math>V</math> = volume of the core in <math>cm^3</math>                      Loop area is in gauss-oersteds.</p>
$P_r$	watts	Residual Loss	<p>Part of the total core losses which is due neither to <math>P_e</math> nor to <math>P_h</math>. It is the most important source of losses in ferrites. The residual losses cannot be precisely calculated, but should be measured.</p>
$P_\Delta$	watts	Incremental Core Loss	<p>The core loss in a ferromagnetic material when subjected simultaneously to a biasing and an incremental magnetizing force.</p>
$P_a$	watts	Apparent Core Loss	<p>The product of the R.M.S. induced e.m.f. and R.M.S. magnetizing current for a coil containing a ferromagnetic core. The induced e.m.f. must be approximately sinusoidal.</p>

SYMBOL	UNIT	TERM	DESCRIPTION
$P_{cu}$	Watts	Copper Loss	Power dissipated in the coil windings due to the electrical resistance of the copper wire. The resistance increases with the frequency due to the so-called skin effect, when the current flows mainly through the outmost part of the wire. These losses are considerable in pot core applications at high frequencies and have a large influence on the maximum attainable Q. These losses can be reduced by using Litz wire.
Q	numeric	Quality(Factor)	$Q = \frac{\omega L}{R_s}$ <p>where: <math>R_s</math> represents the effective series-resistance in ohms in which the magnetizing current dissipates a power equal to total core losses,  <math>L</math> = pure self-inductance of the core in henries,  <math>\omega = 2\pi f</math> = angular frequency at which the core resonates with a series or parallel capacitor. The smaller the losses, the larger the Q. Q usually decreases with increasing frequency.</p> <p>At low inductions we can also write: <math>Q = \frac{\mu^3}{\mu^2}</math></p>
$\delta$		Loss Angle	$\frac{1}{Q} = \frac{R_s}{\omega L} = \tan \delta$ <p>where: <math>\delta</math> = the loss-angle.  <math>\delta</math> is actually the phase angle with which the magnetizing current lags the induced e.m.f. across the coil at low inductions.</p>
$\frac{1}{\mu_o Q}$		Loss Factor	<p>The loss factor is defined as <math>\frac{1}{\mu_o Q} = \frac{\tan \delta}{\mu_o} = \left( \frac{\mu^2}{\mu_o^2} \right)</math>.</p> <p><math>\frac{1}{\mu_o Q}</math> is an important characteristic for both the losses and the <math>\mu_o</math>.</p>
		Core Material Losses	<p>A quantity which is of interest for losses at low inductions. It is written: (compare with <math>\frac{1}{\mu_o Q} = \frac{R_s}{\mu_o^2 \pi f L}</math>) as</p> $\frac{R_s}{\mu_o f L}$ <p>where: <math>R_s</math> = effective series resistance,  <math>\mu_o</math> = initial permeability,  <math>f</math> = frequency in cycles/sec,  <math>L</math> = inductance in henries.</p> <p>The formula can be divided into three parts:</p> $\frac{R}{\mu_o f L} = C_h B_{max.} + C_e f + C_r$ <p>where: <math>C_h</math>, <math>C_e</math> and <math>C_r</math> are the hysteresis, eddy current and residual loss coefficients.</p>

8. MEMORY CORES

MEMORY CORE

A small size toroid of square loop material. If when operated with an a.c. magnetic field with peak values greater than the coercivity  $H_{cs}$ , the induction jumps between the positive,  $+B_s$ , and the negative,  $-B_s$ , saturation induction, then the core is a ferromagnetic switch. Therefore, when the core receives a sufficiently strong positive magnetizing current pulse, it will jump into  $+B_s$  and stay in its positive retentivity,  $+B_{rs}$ : it "remembers" the positive pulse.  $-B_{rs}$  might be called the "0" or zero state and  $+B_{rs}$  the "1" state of the core. Memory cores are used in digital computers as storage elements of information or in shift registers. See Fig. 6.

PULSES

A memory core is tested with a succession of magnetizing current pulses of which the repetition frequency, peak values and individual forms are specified by the customer. The pulses are applied via a single wire or probe through the center-hole of the core and cause induced voltages. Their limits are also specified. The same succession is repeated for each core. See Fig. 7 and 9.

PULSE SEQUENCE

Succession of pulses, whether write ("1") or read ("0") and whether full or partial. This determines the change of induction (flux) per second,  $\frac{d\phi}{dt}$ , and therefore the response ("answer" of the core in terms of the induced e.m.f. in the probe) on a pulse. See Fig. 9.

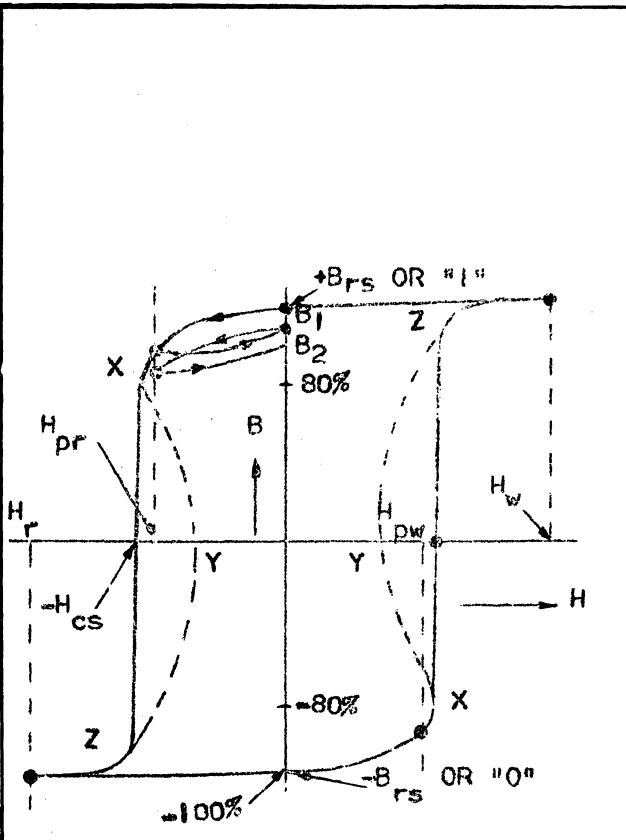


FIGURE 6

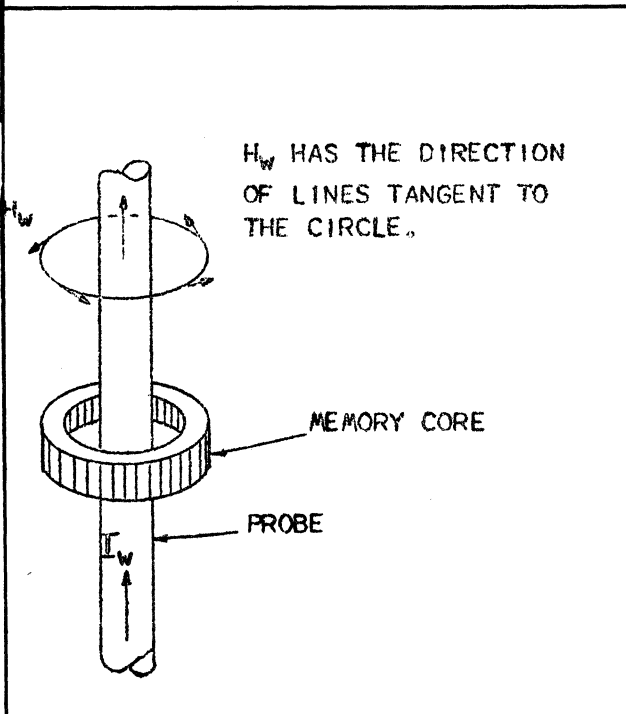


FIGURE 7

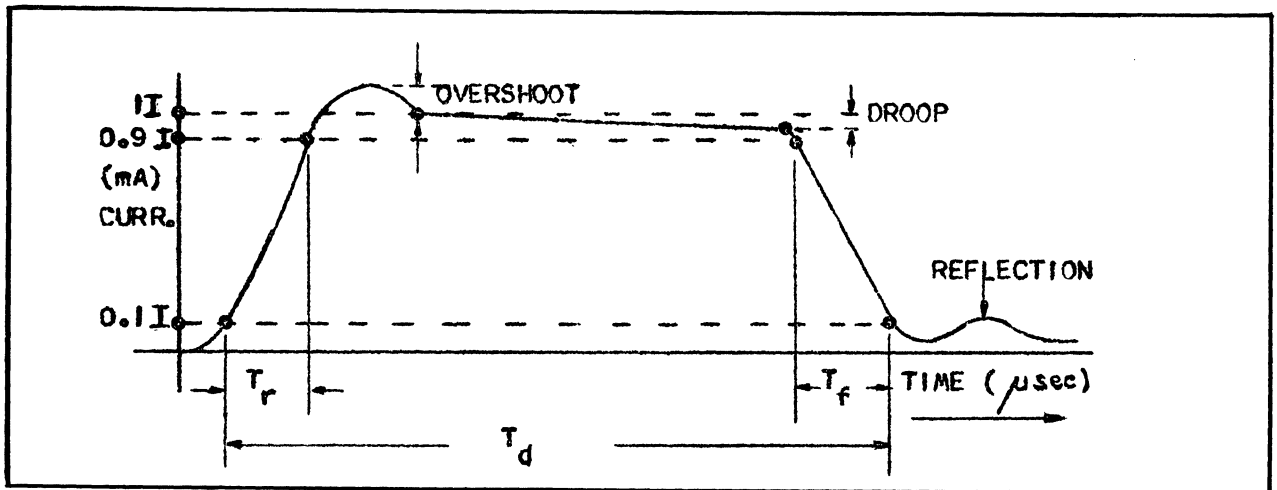


FIGURE 8

THE GENERAL SHAPE OF THE PULSES USED

SYMBOL	UNIT	TERM	DESCRIPTION
$I$	mA	Peak Current	Peak value ( $I$ ) of current pulse in milliamperes, See Fig. 8. The drawing is exaggerated for clarity.
$T_r$	$\mu$ sec	Rise Time	Time for the pulse current to rise from $0.1 I$ to $0.9 I$ . See Fig. 8.
$T_d$	$\mu$ sec	Duration	Time for the pulse current to stay above $0.9 I$ . See Fig. 8.
$T_f$	$\mu$ sec	Fall Time	Time for the pulse current to drop from $0.9 I$ down to $0.1 I$ . See Fig. 8.

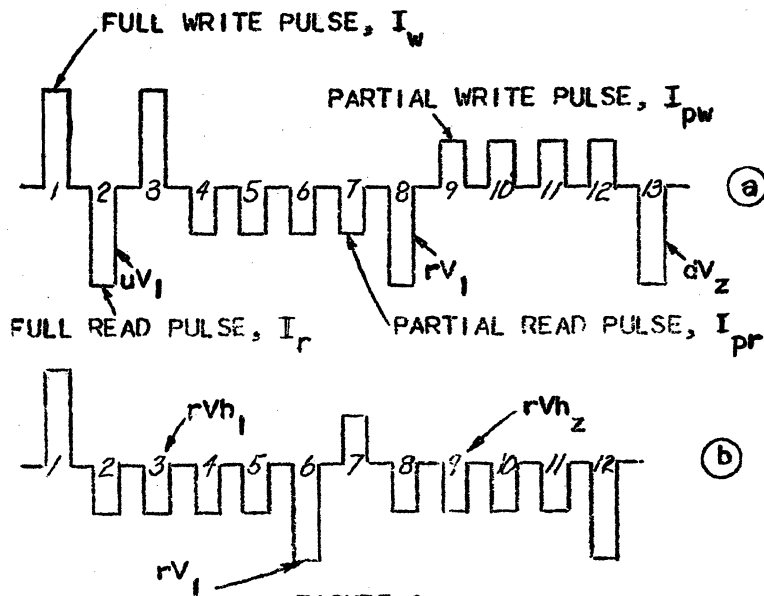


FIGURE 9

TWO CHARACTERISTIC PULSE SEQUENCES USED IN TESTS

SYMBOL	UNIT	TERM	DESCRIPTION
$I_w$	mA	Full Write Pulse	Peak value of pulse, that brings core into the "1" state or $+B_{rs}$ . The magnetizing force corresponding to $I_w$ is $H_w$ . See Figs. 6 and 7.
$I_{pw}$	mA	Partial Write Pulse	Peak value of pulse, that does not manage to bring core from the "0" to the "1", but that changes "0" only to a slightly different value (disturbs it). The magnetizing force corresponding to $I_{pw}$ is $H_{pw}$ . See Figs. 6 and 7.
$I_r$	mA	Full Read Pulse	Peak value of negative pulse, that brings core into "0" from "1". In order to explain the form of $uV_1$ or $rV_1$ we must assume that not the static hysteresis loop, but a loop with curved sides (Fig. 6 XYZ) is performed by B and H.
$I_{pr}$	mA	Partial Read Pulse	Peak value of negative pulse, that disturbs the "1" state of the core. In upper left of figure 6, $B_1$ and $B_2$ indicate the residual induction after one and two $I_{pr}$ 's respectively. If more $I_{pr}$ 's are applied, the final residual induction will be very little different from $B_1$ or $B_2$ .
$I_0$	mA	Threshold Current	The peak value of the current pulse, whether positive or negative, that brings the residual induction, $B_r$ , a certain amount below the retentivity, $B_{rs}$ , say 20%. These values are indicated in Fig. 6.



SYMBOL	UNIT	TERM	DESCRIPTION
$uV_1$	mV (peak value)	Undisturbed Voltage One	Response (see fig. 10) of core measured when an $I_w$ is followed immediately by and $I_r$ . See fig. 9a: 1→2.
$rV_1$ or $dV_1$	mV (peak value)	Disturbed Voltage One	Peak response of core measured, when an $I_w$ is disturbed by a certain number of $I_{pr}$ 's followed by an $I_r$ . See fig. 9a: 3→4--→7 → 8 or fig. 9b: 1→ 2---→5→6.
$dV_2$	mV (peak value)	Disturbed Voltage Zero	Peak response, when an $I_r$ is disturbed by a certain number of $I_{pw}$ 's followed by an $I_r$ . See fig. 9a: 8→9--→12 → 13.
$rVh_1$	mV (peak value)	Disturbed Half- Selected One	Response of the <u>second</u> $I_{pr}$ after an $I_w$ . See fig. 9b, 1→ 2→ 3
$rVh_2$	mV (peak value)	Disturbed Half- Selected Zero	Response of the <u>second</u> $I_{pr}$ after an $I_{pw}$ . See fig. 9b: 7→ 8 → 9.
$T_p$	$\mu$ sec	Peaking Time	Time for $rV_1$ to rise from reference voltage level to its peak-value. See fig. 10.
$T_s$	$\mu$ sec	Switching Time	Time for $rV_1$ to rise from and drop back to the reference voltage level. See fig. 10.

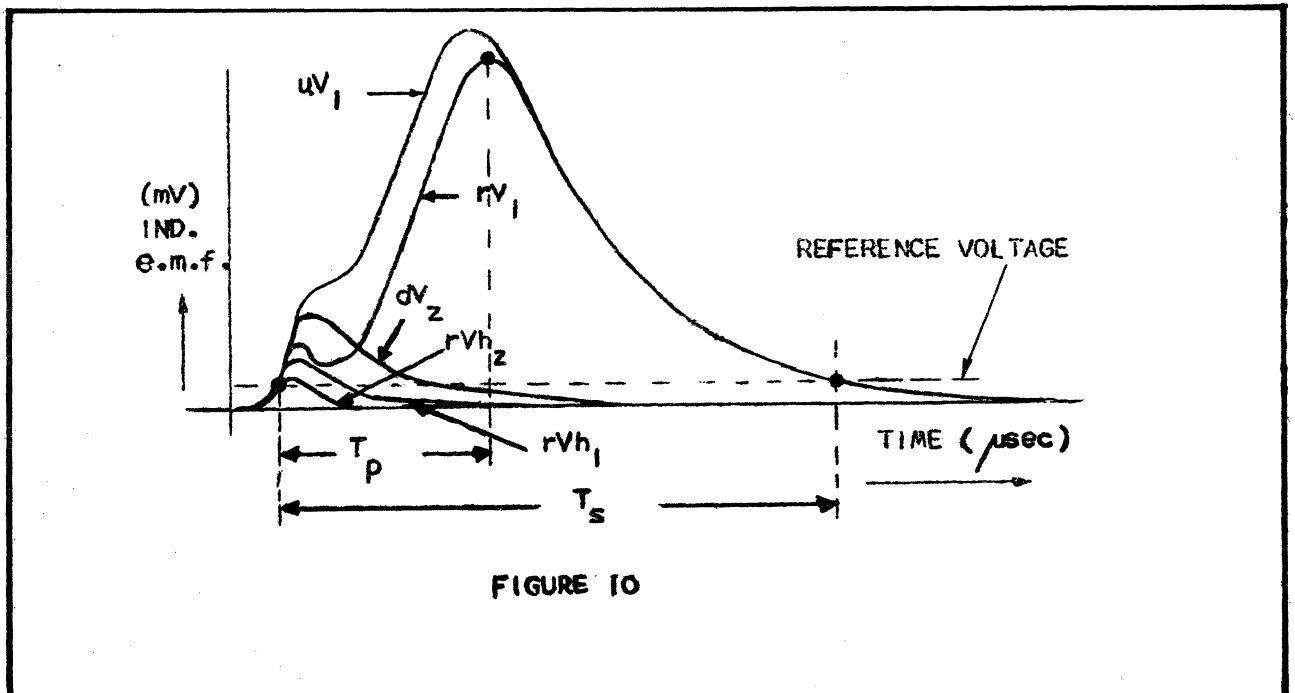


FIGURE 10

9. MISCELLANEOUS

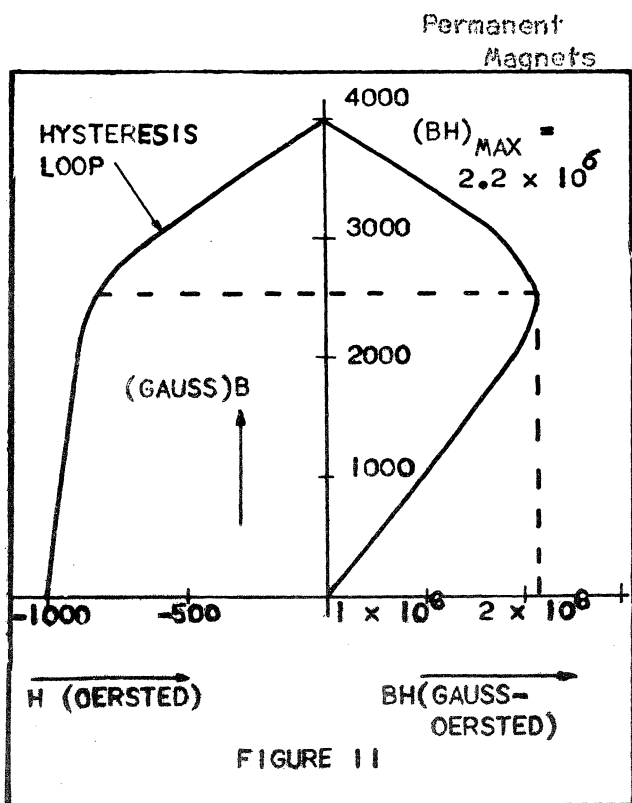
SYMBOL	UNIT	TERM	DESCRIPTION
$T_c$	$^{\circ}C$	Curie Temperature	<p>Temperature in degrees centigrade at which a material becomes non-magnetic, actually indistinguishable from air (<math>\mu_o = 1</math>).</p> <p>Generally the <math>\mu_o</math> will increase with rising temperature but will decrease quickly to 1 at the Curie temperature. Further, the higher the <math>\mu_o</math>, the lower the <math>T_c</math> will be. If for ferrites <math>\mu_o \approx 10</math>, then <math>T_c</math> is about <math>500^{\circ}C</math>; if <math>\mu_o \approx 1000</math>, then <math>T_c</math> is about <math>150^{\circ}C</math>.</p>
T.C.	$\frac{\text{percent}}{^{\circ}C}$	Temperature Coefficient	<p>The relative change of a quantity Y (e.g. <math>\mu_o</math>, <math>\mu_{\Delta}</math>, <math>B_s</math>, <math>\frac{\text{Tan } \delta}{\mu_o}</math>) per degree centigrade or:</p> $T.C. = \frac{1}{Y} \times \frac{\Delta Y}{\Delta T}$ <p>where: <math>\Delta Y</math> = change (increase or decrease) in Y occurring at a change in temperature, <math>\Delta T</math>.</p> <p><math>\frac{1}{Y} \times \frac{\Delta Y}{\Delta T} \times 100\%</math> gives the percentage change per degree centigrade.</p> <p>In precision applications a very low T.C. for self inductances is desirable. A negative T.C. for the losses would prevent excessive heating up of core.</p>
$\rho$	$\text{gr/cm}^3$	Density Specific Weight	<p>This is defined as total mass (weight) divided by the volume of a body:</p> $\rho = \frac{\text{total mass}}{\text{total volume}}$ <p>in grams per cubic centimeter.</p>
		Disaccommodation	<p>The decrease of <math>\mu_o</math> immediately following the demagnetization of the core. In the course of 24 hours, the lower limit, say 3% lower, is approached. The same seems to happen with the reversible permeability, <math>\mu_{rev}</math>. This results that, where the biasing field is changed, a ferrite cannot be used for precision applications. Its effect can be reduced by air gaps.</p>

SYMBOL	UNIT	TERM	DESCRIPTION
$\rho$	$\Omega \text{ cm}$	(Volume) Resistivity Specific Resistance	The d.c. resistance of a cylindrical conductor (bar) with a length of 1 cm and a perpendicular cross-section = 1 cm <sup>2</sup> . It is measured between the two cross-sections which are wholly contacted by the measuring electrodes. With the resistivity known, we can calculate the d.c. resistance of a bar with a length $l$ cm and a cross-section $A$ cm <sup>2</sup> from $R = \rho \times \frac{l}{A}$ .
$f$	c/sec	Frequency	The number of times that a periodic quantity (e.g. alternating current) repeats itself during 1 second. It is expressed in cycles per second. 1 Kc/sec. = 1 kilocycle per second = 10 <sup>3</sup> c/sec. $\omega = 2 \pi f$ is called the angular frequency in radians per sec.

$\lambda$  (lambda) numeric Magnetostriction

When a ferromagnetic rod is magnetized by the coil around it, the length of the rod will change by a small amount  $\Delta l$ . The magnetostriction at a given induction  $B$  is defined as  $\lambda = \frac{\Delta l}{l}$ .

$\Delta l$  can also be negative. Magnetostriction can be used to convert mechanical vibrations into electrical.



Hard (high  $H_{cs}$ ) magnetic materials can be used as permanent magnets (Magnadur), as, for example, in moving coil meters, where we need a strong magnetic field in the air-gap for the moving coil. The maximum possible field  $H$  is determined by the maximum value of the energy product,  $(BH)_{max}$ .

The energy product curve is obtained by multiplying  $H$  with  $B$  at every point of the hysteresis-loop where  $B$  is positive and  $H$  is negative and plotting these values against  $B$ . See Fig. 11

SYMBOL	UNIT	TERM	DESCRIPTION
$\mathcal{R}$	1/cm	(Magnetic) Reluctance	<p>The reluctance of a magnetic circuit is defined as:</p> $\mathcal{R} = \frac{\text{Magnetomotive Force}}{\text{Flux}} = \frac{\mathcal{F}}{\phi}$ <p>It can be proven that <math>\mathcal{R} = \frac{l}{\mu A}</math> for circuit with uniform cross-section <math>A(\text{cm}^2)</math>, length <math>l(\text{cm})</math> and a permeability <math>\mu</math>. If not, <math>\frac{l}{\mu A}</math> has to be replaced by the sum of <math>\frac{l}{\mu A}</math> for the different parts of the circuit (e.g. pot core) or generally:</p> $\mathcal{R} = \sum \frac{l}{\mu A}$ <p><math>\mathcal{R}</math> can be expressed in <math>\frac{1}{\text{cm}}</math> or gilbert/maxwell.</p> <p>Note: Compare definition of electrical resistance.</p> $R = \frac{\text{electromotive force}}{\text{current}} = \frac{V}{I} = \sum \frac{l}{A}$

10. UNITS IN THE CENTIMETER-GRAM-SECOND (c.g.s.) SYSTEM AND THE METER-KILOGRAM-SECOND (m.k.s.) SYSTEM

UNIT	ABBREV.	QUALITY	DESCRIPTION
Mega	M	$\times 10^6$	<p>Mega = times one million = <math>\times 10^6</math> written before a unit, it means that the unit is multiplied by <math>10^6</math> and thus has become a more practical unit depending on the application.</p> <p>(e.g. <math>3.2 \text{ M}\Omega = 3.2 \text{ megohms} = 3.2 \times 10^6 \Omega</math>;  <math>7 \text{ MV} = 7 \text{ megavolts} = 7 \times 10^6 \text{V}</math>).</p>
Kilo	K	$\times 10^3$	<p>Kilo = times one thousand = <math>\times 10^3</math>.</p> <p>8 KG = 8 kilograms = 8000 g (grams).</p>
Milli	m	$\times 10^{-3}$	<p>Milli = times one thousandth = <math>\times 10^{-3}</math>.</p> <p>1 mV = 1 millivolt = <math>10^{-3} \text{V}</math>.</p>
Micro	$\mu$	$\times 10^{-6}$	<p>Micro = times one millionth = <math>\times 10^{-6}</math>.</p> <p>9 <math>\mu\text{V}</math> = 9 microvolts = <math>9 \times 10^{-6} \text{V}</math>.</p>
Pico or Micro-micro	p $\mu\mu$	$\times 10^{-12}$	<p>Pico Micro-micro } = <math>\times 10^{-12}</math></p> <p>1 <math>\mu\mu\text{A}</math> = 1 micro-micro ampere = <math>10^{-12} \text{A}</math> (amperes).</p> <p>Usually one writes 1 mA instead of <math>\mu\mu\text{A}</math> which is actually incorrect.</p>

UNIT	ABBREV.	QUALITY	DESCRIPTION
Centimeter	cm	Length	Unit of length, originally defined as a certain fraction of the earth's circumference. Consequently we have
Square Centimeter	cm <sup>2</sup>	Area	1 cm <sup>2</sup> and 1 cm <sup>3</sup> as units of area and volume respectively. 1 inch = 2.54 cm; 1 square inch = 6.45 cm <sup>2</sup> ; 1 cubic inch = 16.4 cm <sup>3</sup> . M.k.s. system: 1 meter = 100 cm; 1 m <sup>2</sup> = 10 <sup>4</sup> cm <sup>2</sup> ; 1 m <sup>3</sup> = 10 <sup>6</sup> cm <sup>3</sup> .
Cubic Centimeter	cm <sup>3</sup>	Volume	
Ampere	A	Current	Unit of current defined as the unit of positive electrical charge per second. The same unit is used in the m.k.s. system.
Erg		Work	See Watt
Farad	F	Capacitance	Unit of capacitance, defined as the constant ratio between electrical charge on, and voltage between, two neighboring conductors (plates of a condenser). Same in m.k.s.:
			$1 \mu F = 10^{-6} F.$
Gauss	G	Induction	The unit of magnetic induction or magnetic flux density. Definition: See "Induction". In the m.k.s. system; the unit of induction is 1 Weber per square meter or 1 voltsecond per square meter. $1 \text{ weber/m}^2 = 1 \text{ voltsec/m}^2 = 10^4 \text{ gauss.}$
Gilbert		Magnetomotive Force	The unit of magnetomotive force. It is seldom used.
Gram	g gr	Mass (Weight)	Unit of mass. Kilogram and milligram are also much used. M.k.s. system: 1 kilogram = 1000 grams.
Henry	H	Self Inductance Mutual Inductance	The unit of self inductance and mutual inductance. Definition: see under "Inductance". M.k.s. system: same unit. 1 mH = 1 millihenry = 10 <sup>-3</sup> H.
Maxwell		Flux	The units of flux. M.k.s. system: 1 weber (= 1 voltsec) = 10 <sup>8</sup> maxwells. Description: see under "Flux". (In the m.k.s. system the formula of induced e.m.f. is;
			$e_{ind} = - \frac{d(N\phi)}{dt}$
			where:
			$N\phi$ = flux linkage in weber-turns,
			$e_{ind}$ = induced voltage in volts.

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UNIT	ABBREV.	QUALITY	DESCRIPTION
Ohm	$\Omega$	Resistance	<p>Unit of resistance, defined as :</p> $R = \frac{V}{I}$ <p>where:</p> <p>V = voltage across conductor with the resistance in V,                      I = current in A,                      R = resistance in ohms.                      M.k.s. system same unit.</p>
Oersted	Oe	Magnetizing Force	<p>Unit of magnetizing force. Definition: see under "Magnetic Field". M.k.s. system; 1 ampere turn per meter = 1 At/m = 40 <math>\pi</math> Oe = <math>1.257 \times 10^2</math> Oe. In the m.k.s. system the formula for H in a homogeneously wound toroid is:</p> $H = \frac{NI}{l}$ <p>where:</p> <p>N = number of turns,                      I = current in A                      l = pathlength in m,                      H = in At/m.</p>
Second	sec	Time	<p>Unit of time in both systems. Defined as a certain fraction of the time of revolution of the earth around the sun.</p>
Volt	V	Electromotive Force	<p>Unit of voltage. Can be defined as work per unit charge.</p>
Watt	W Joule/sec	Power Work per sec. Energy per sec. (Heat dissipated per sec.)	<p>Unit of power. The power of a motor is the work it can achieve in one second. Power = <math>\frac{\text{Work}}{\text{Second}}</math>. Unit of work is the erg, therefore unit of power = 1 erg/sec. One usually uses the watt or joule/sec from the m.k.s. system: 1 watt = <math>10^7</math> ergs/sec.</p>





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